

FEATURES

Downconverter

Conversion loss

8 dB typical for 3 GHz to 7 GHz

10 dB typical for 7 GHz to 10 GHz

LO to IF isolation

44 dB typical for 3 GHz to 7 GHz

42 dB typical for 7 GHz to 10 GHz

RF to IF isolation

21 dB typical for 3 GHz to 7 GHz

32 dB typical for 7 GHz to 10 GHz

Input IP3

22 dBm typical for 3 GHz to 7 GHz

28 dBm typical for 7 GHz to 10 GHz

Input P1dB

15 dBm typical for 3 GHz to 7 GHz

17 dBm typical for 7 GHz to 10 GHz

Passive double balanced topology

Wide IF frequency range: dc to 4 GHz

6-pad, bare die [CHIP]

APPLICATIONS

Microwave radio

Industrial, scientific, and medical (ISM) band and ultrawide band (UWB) radio

Test equipment and sensors

Military end use

GENERAL DESCRIPTION

The [HMC787AG](#) is a general-purpose, double balanced mixer that can be used as an upconverter or downconverter from 3 GHz to 10 GHz. This mixer is fabricated in a gallium arsenide (GaAs), metal semiconductor field effect transistor (MESFET) process and requires no external components or matching circuitry.

FUNCTIONAL BLOCK DIAGRAM

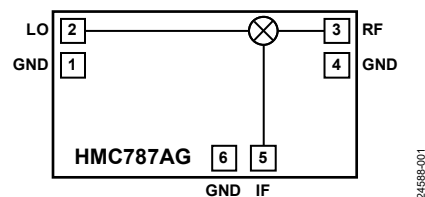


Figure 1.

24588-001

The HMC787AG provides high local oscillator (LO) to RF and LO to intermediate frequency (IF) isolation due to optimized balun structures and operates with a LO drive level between 13 dBm to 21 dBm.

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REVISION HISTORY

3/2021—Revision 0: Initial Version

SPECIFICATIONS

ELECTRICAL SPECIFICATIONS

$T_A = 25^\circ\text{C}$, IF = 100 MHz, LO drive level = 17 dBm, and all measurements performed as a downconverter, unless otherwise noted.

Table 1.

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
FREQUENCY RANGE					
RF		3		10	GHz
IF		DC		4	GHz
LO		3		10	GHz
LO DRIVE LEVEL		13	17	21	dBm
3 GHz TO 7 GHz PERFORMANCE					
Downconverter					
Conversion Loss			8	10	dB
Single Sideband Noise Figure	Measurements taken with external LO amplifier		8		dB
Input Third-Order Intercept (IP3)	1 MHz separation between inputs	16	22		dBm
Input 1 dB Compression Point (P1dB)			15		dBm
Input Second-Order Intercept (IP2)	1 MHz separation between inputs		74		dBm
Upconverter					
Conversion Loss			8		dB
Input IP3	1 MHz separation between inputs		37		dBm
Isolation					
RF to IF		15	21		dB
LO to RF			47		dB
LO to IF		36	44		dB
7 GHz TO 10 GHz PERFORMANCE					
Downconverter					
Conversion Loss			10	11	dB
Single Sideband Noise Figure	Measurements taken with external LO amplifier		10		dB
Input IP3	1 MHz separation between inputs	22	28		dBm
Input P1dB			17		dBm
Input IP2	1 MHz separation between inputs		70		dBm
Upconverter					
Conversion Loss			9		dB
Input IP3	1 MHz separation between inputs		36		dBm
Isolation					
RF to IF		24	32		dB
LO to RF			43		dB
LO to IF		25	42		dB

ABSOLUTE MAXIMUM RATINGS

Table 2.

Parameter	Rating
Input Power	
RF	28 dBm
LO	28 dBm
IF	28 dBm
IF Source and Sink Current	12 mA
Continuous Power Dissipation, P_{DISS} ($T_A = 85^\circ\text{C}$, Derate 11.6 mW/ $^\circ\text{C}$ Above 85°C)	1044 mW
Temperature	
Maximum Junction	175°C
Operating Range	-40°C to $+85^\circ\text{C}$
Storage Range	-65°C to $+150^\circ\text{C}$

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

ELECTROSTATIC DISCHARGE (ESD) RATINGS

The following ESD information is provided for handling of ESD sensitive devices in an ESD protected area only.

Human body model (HBM) per ANSI/ESDA/JEDEC JS-001.

Field induced charged device model (FICDM) per ANSI/ESDA/JEDEC JS-002.

ESD Ratings for HMC787AG

Table 3. HMC787AG, 6-Pad CHIP

ESD Model	Withstand Threshold (V)	Class
HBM	350	1A
FICDM	1250	C3

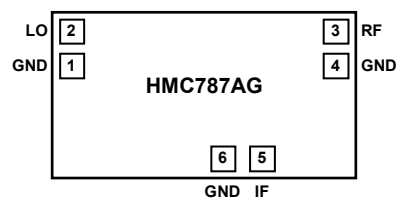
ESD CAUTION



ESD (electrostatic discharge) sensitive device.

Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



NOTES
1. DIE BOTTOM. CONNECT THE DIE BOTTOM TO RF AND DC GROUND.

Figure 2. Pad Configuration

24588-002

Table 4. Pad Function Descriptions

Pad No.	Mnemonic	Description
1, 4, 6	GND	Ground. See Figure 3 for the GND interface schematic.
2	LO	Local Oscillator. The LO pad is dc-coupled and matched to 50 Ω. See Figure 4 for the LO interface schematic.
3	RF	Radio Frequency. The RF pad is dc-coupled and matched to 50 Ω. See Figure 5 for the RF interface schematic.
5	IF	Intermediate Frequency. The IF pad is dc-coupled. For applications not requiring operation to dc, externally block the IF pad using a series capacitor whose value is chosen to pass the necessary IF frequency range. For operation to dc, the IF pad must not source or sink more than 12 mA of current or device nonfunction and possible device failure results. See for Figure 6 the IF interface schematic.
		Die Bottom. Connect the die bottom to RF and dc ground.

INTERFACE SCHEMATICS



Figure 3. GND Interface Schematic

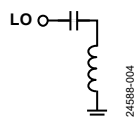


Figure 4. LO Interface Schematic

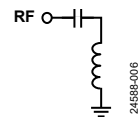


Figure 5. RF Interface Schematic

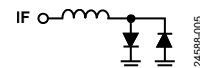


Figure 6. IF Interface Schematic

TYPICAL PERFORMANCE CHARACTERISTICS

DOWNCONVERTER PERFORMANCE, IF = 100 MHz

Lower Sideband (High-Side LO)

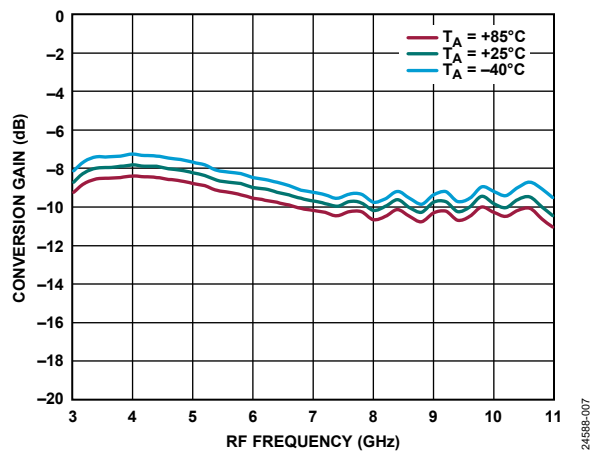


Figure 7. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 17 dBm

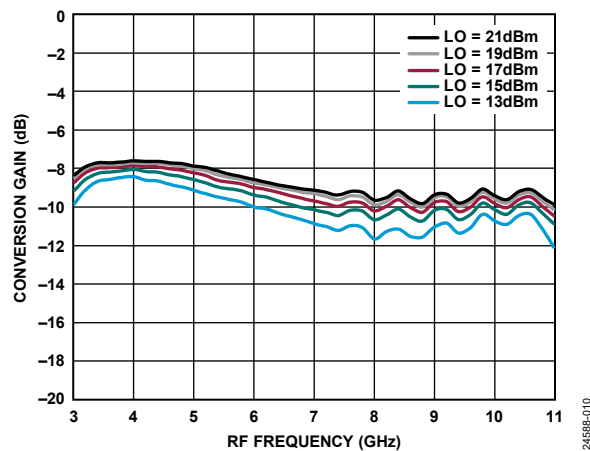


Figure 10. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

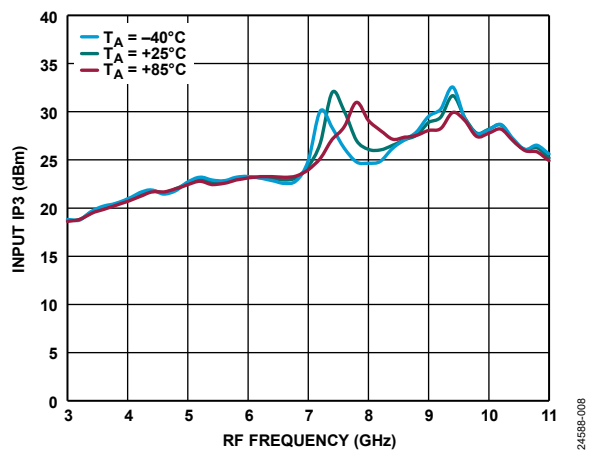


Figure 8. Input IP3 vs. RF Frequency at Various Temperatures, LO = 17 dBm

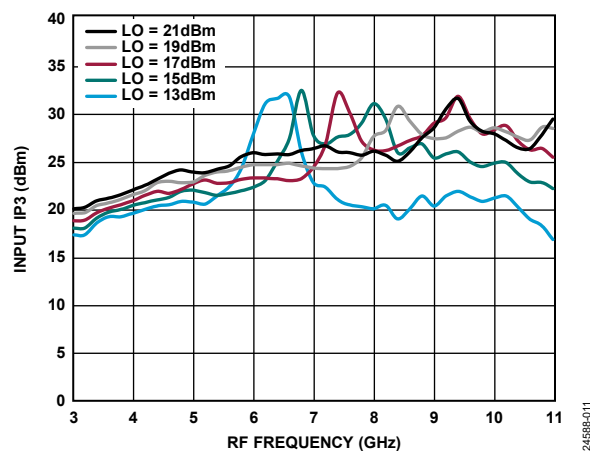


Figure 11. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

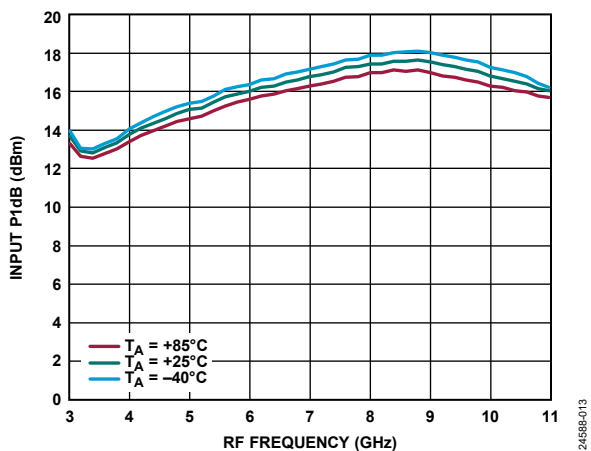


Figure 9. Input P1dB vs. RF Frequency at Various Temperatures, LO = 17 dBm

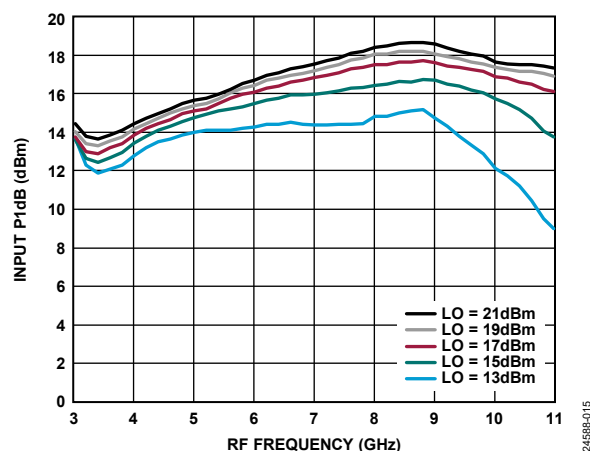


Figure 12. Input P1dB vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

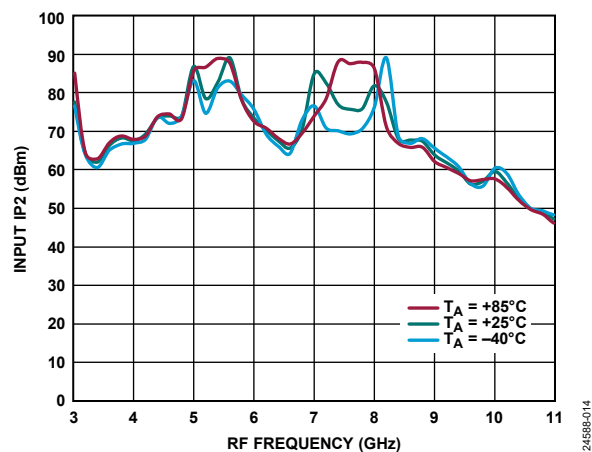


Figure 13. Input IP2 vs. RF Frequency at Various Temperatures,
LO = 17 dBm

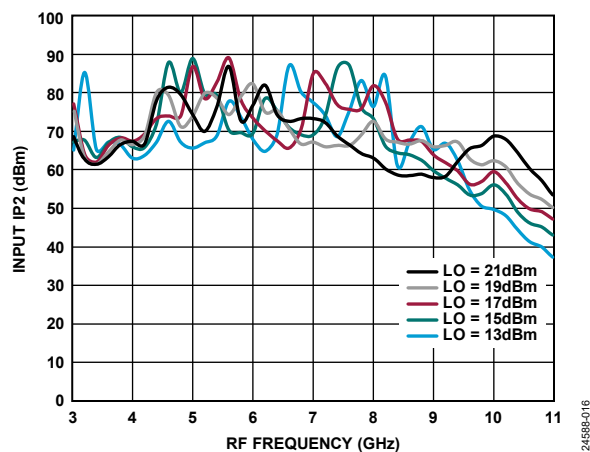


Figure 14. Input IP2 vs. RF Frequency at Various LO Power Levels,
 $T_A = 25^\circ\text{C}$

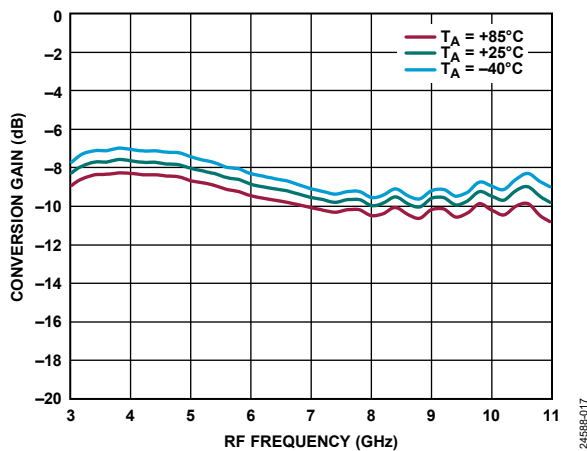
Upper Sideband (Low-Side LO)

Figure 15. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 17 dBm

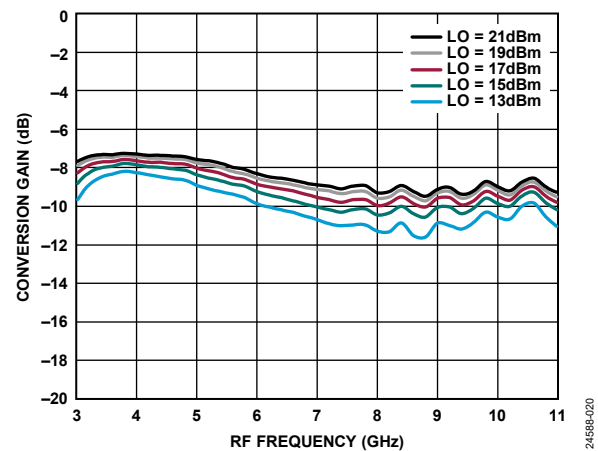


Figure 18. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

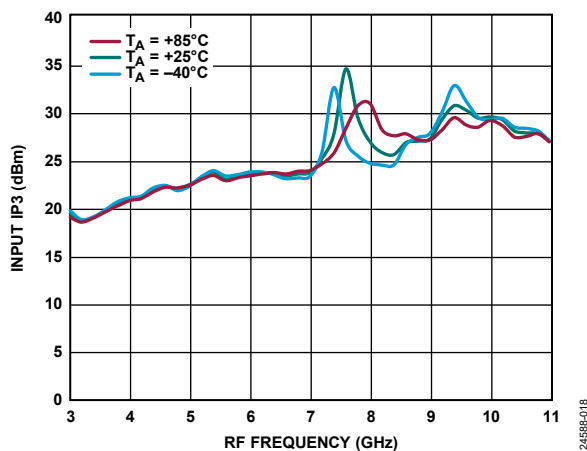


Figure 16. Input IP3 vs. RF Frequency at Various Temperatures, LO = 17 dBm

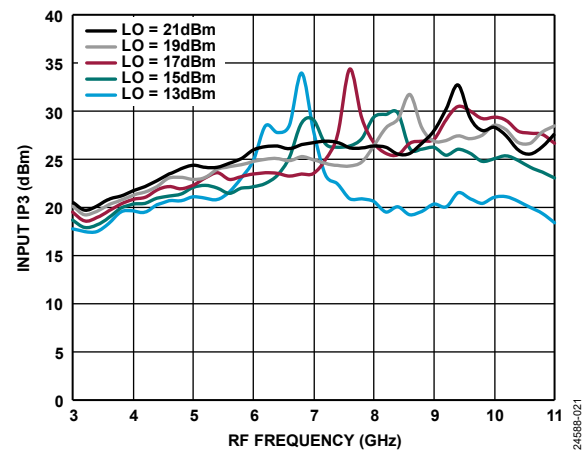


Figure 19. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

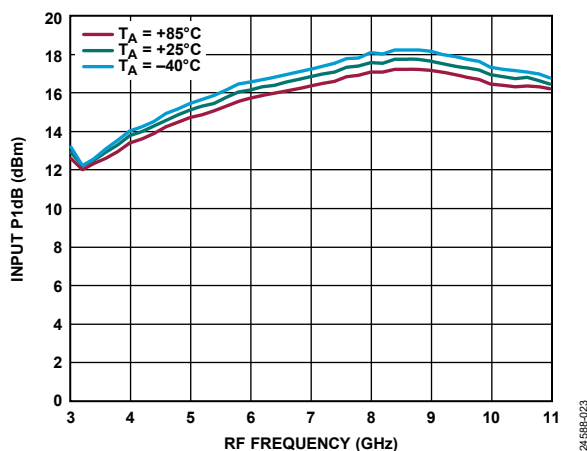


Figure 17. Input P1dB vs. RF Frequency at Various Temperatures, LO = 17 dBm

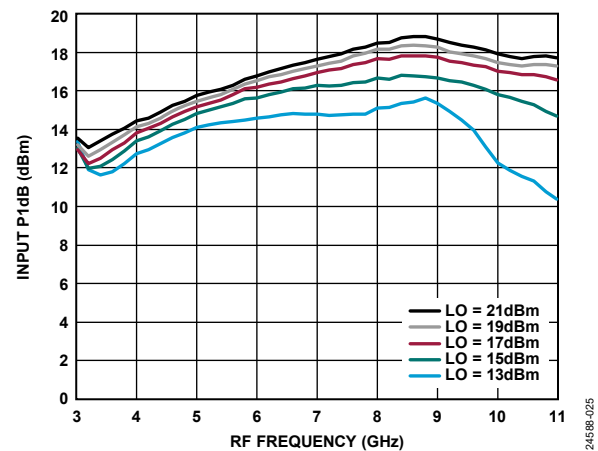


Figure 20. Input P1dB vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

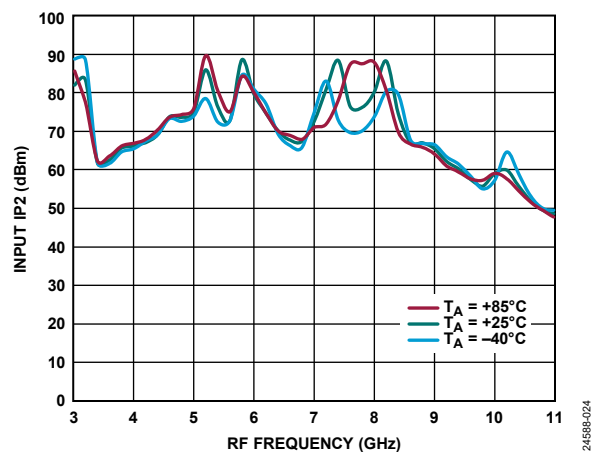


Figure 21. Input IP2 vs. RF Frequency at Various Temperatures,
LO = 17 dBm

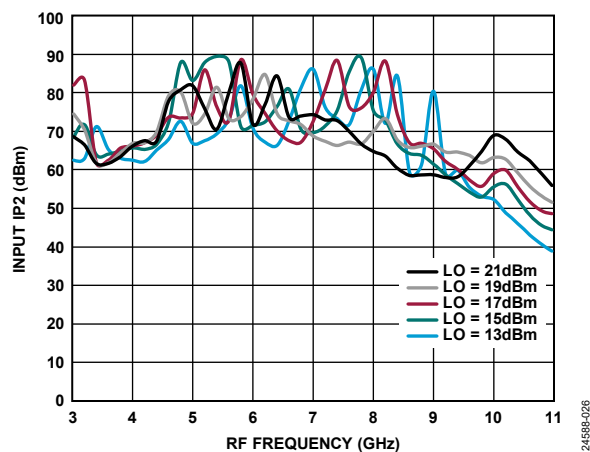


Figure 22. Input IP2 vs. RF Frequency at Various LO Power Levels,
 $T_A = 25^\circ\text{C}$

DOWNCONVERTER PERFORMANCE, IF = 1.1 GHz

Lower Sideband (High-Side LO)

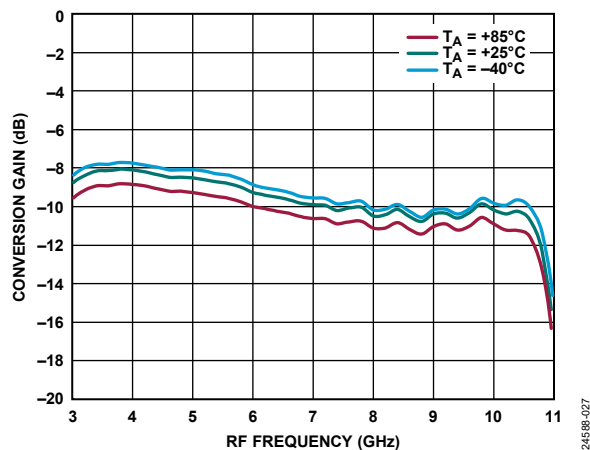


Figure 23. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 17 dBm

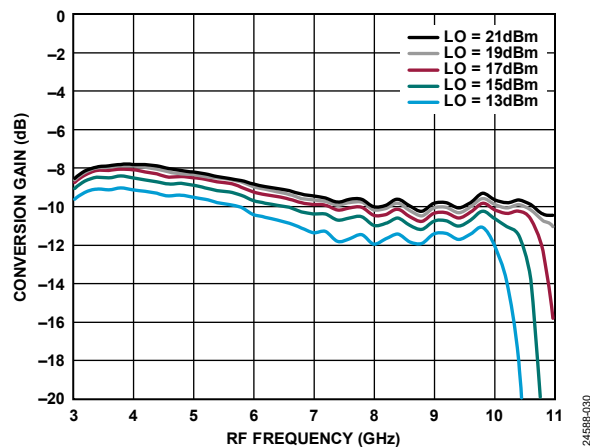
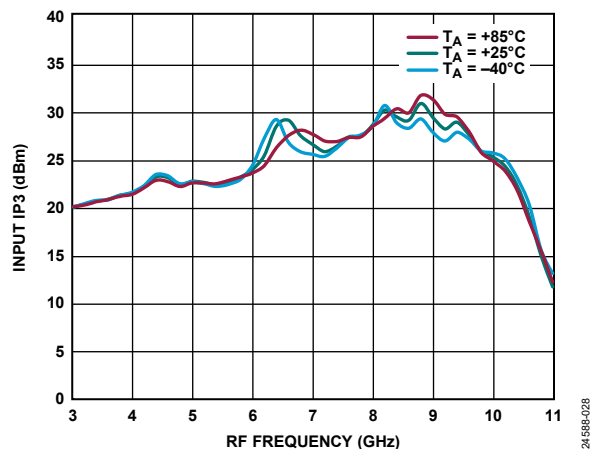
Figure 26. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$ 

Figure 24. Input IP3 vs. RF Frequency at Various Temperatures, LO = 17 dBm

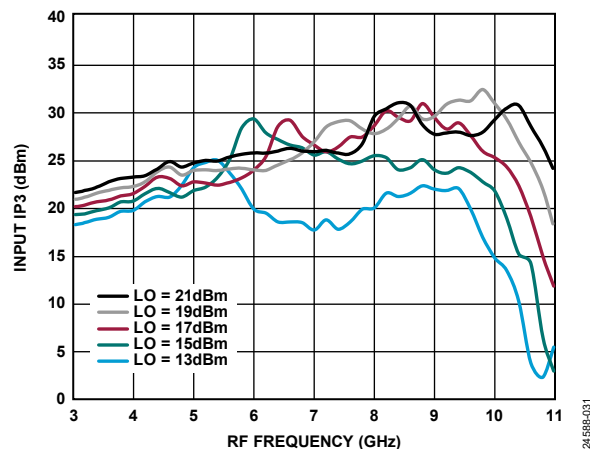
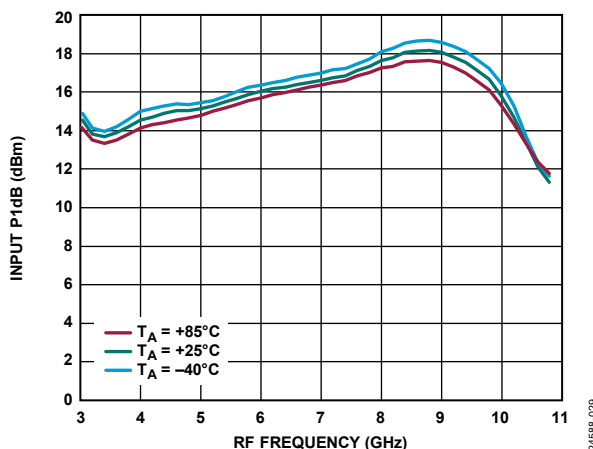
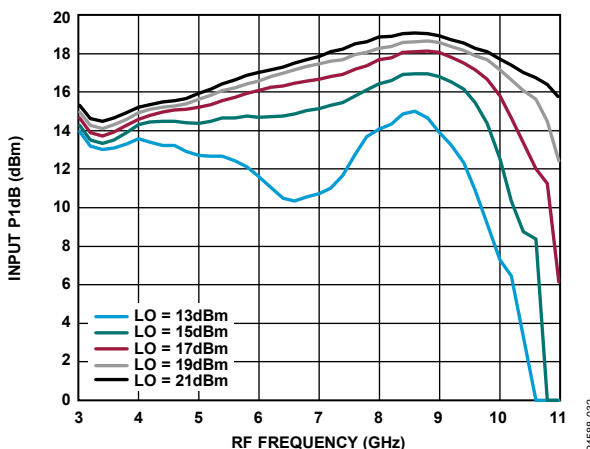
Figure 27. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$ 

Figure 25. Input P1dB vs. RF Frequency at Various Temperatures, LO = 17 dBm

Figure 28. Input P1dB vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

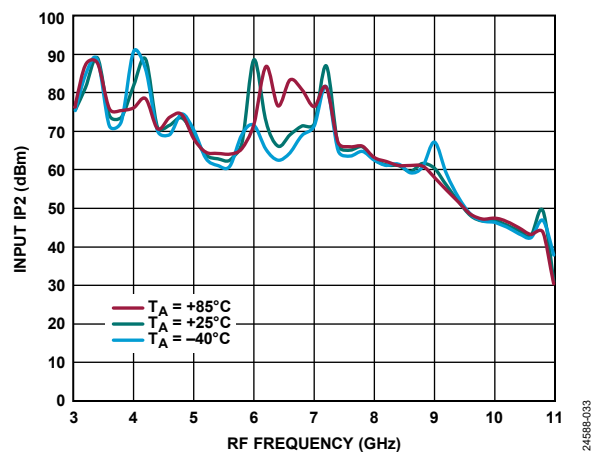


Figure 29. Input IP2 vs. RF Frequency at Various Temperatures,
LO = 17 dBm

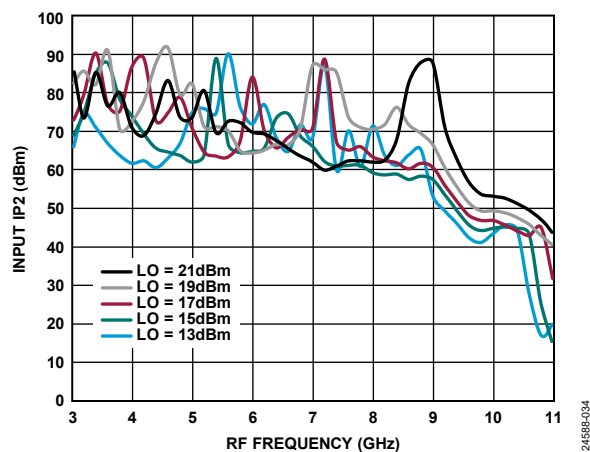


Figure 30. Input IP2 vs. RF Frequency at Various LO Power Levels,
 $T_A = 25^\circ\text{C}$

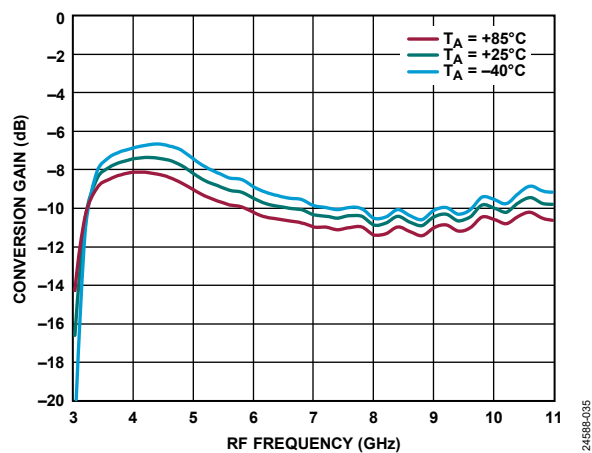
Upper Sideband (Low-Side LO)

Figure 31. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 17 dBm

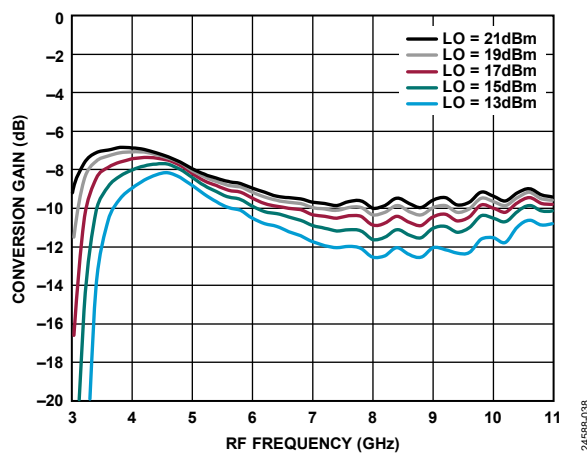


Figure 34. Conversion Gain vs. RF Frequency at Various LO Power Levels, TA = 25°C

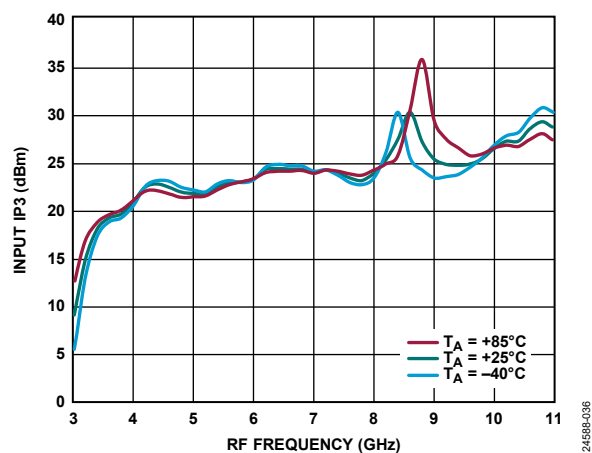


Figure 32. Input IP3 vs. RF Frequency at Various Temperatures, LO = 17 dBm

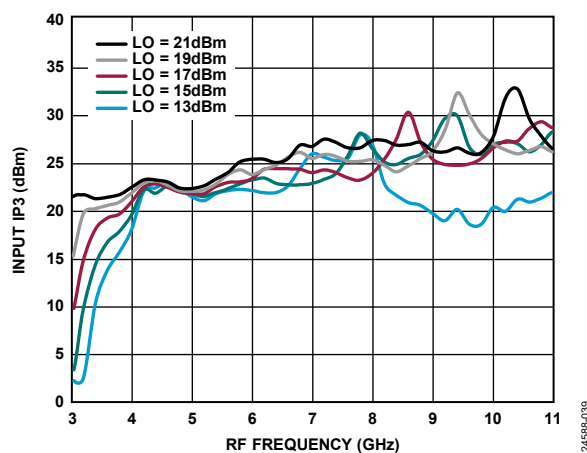


Figure 35. Input IP3 vs. RF Frequency at Various LO Power Levels, TA = 25°C

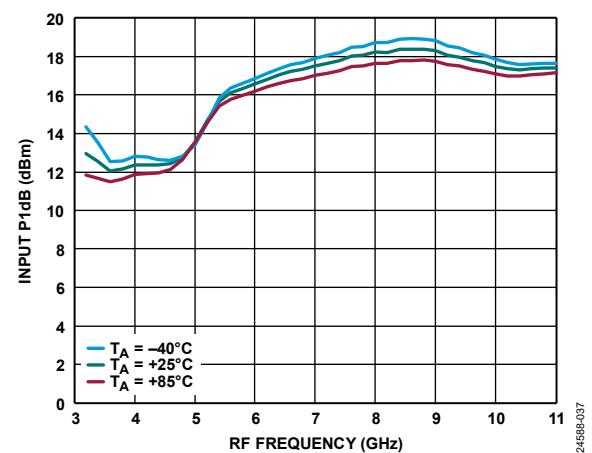


Figure 33. Input P1dB vs. RF Frequency at Various Temperatures, LO = 17 dBm

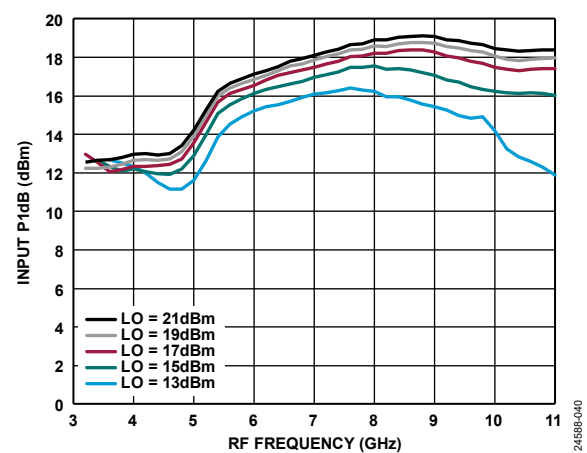


Figure 36. Input P1dB vs. RF Frequency at Various LO Power Levels, TA = 25°C

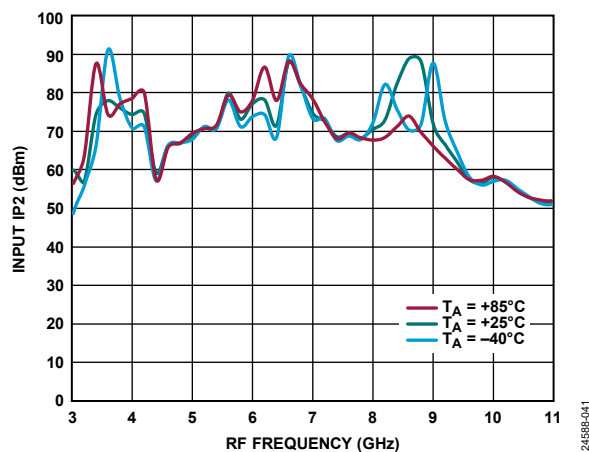


Figure 37. Input IP2 vs. RF Frequency at Various Temperatures
LO = 17 dBm

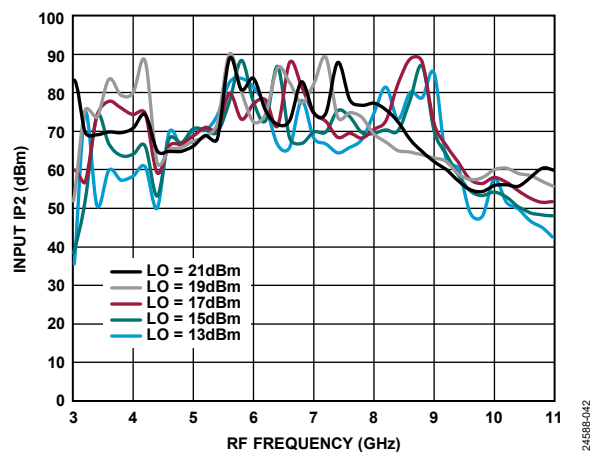


Figure 38. Input IP2 vs. RF Frequency at Various LO Power Levels,
 $T_A = 25^\circ\text{C}$

DOWNCONVERTER PERFORMANCE, IF = 3 GHz

Lower Sideband (High-Side LO)

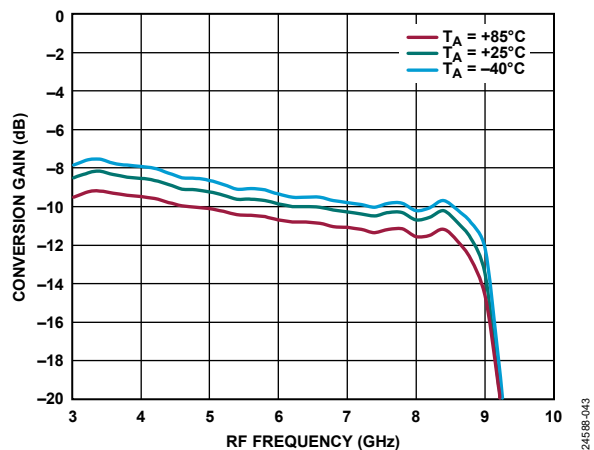


Figure 39. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 17 dBm

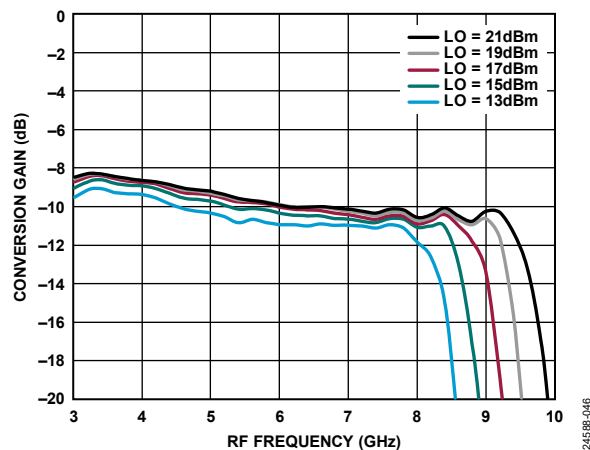
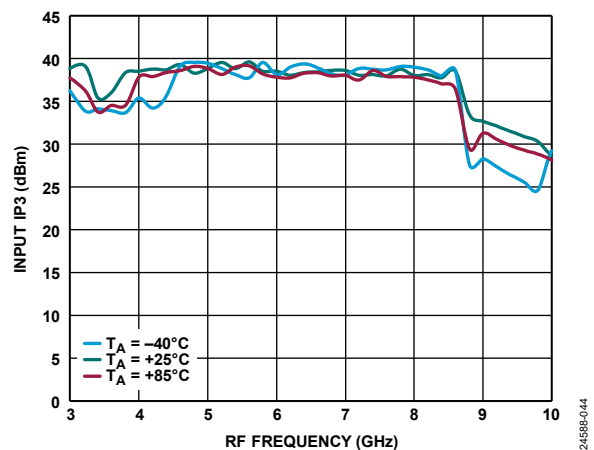
Figure 42. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$ 

Figure 40. Input IP3 vs. RF Frequency at Various Temperatures, LO = 17 dBm

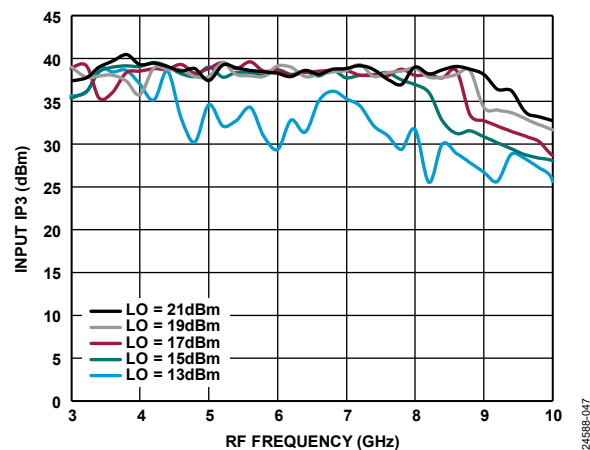
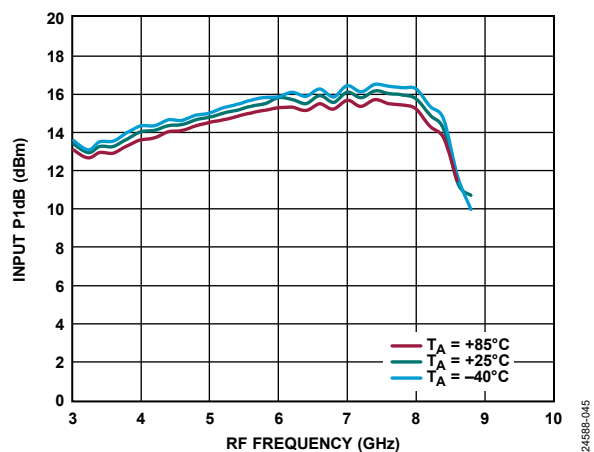
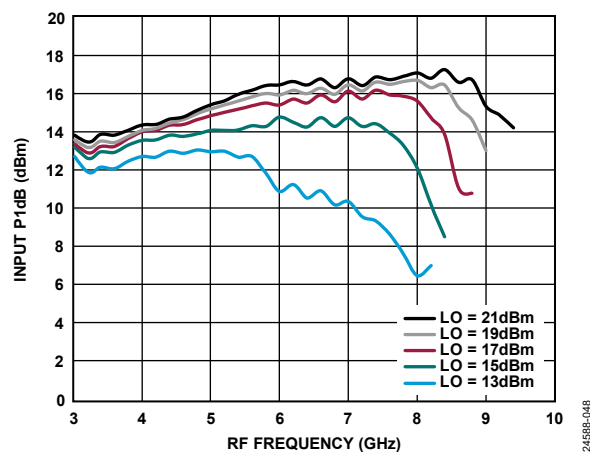
Figure 43. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$ 

Figure 41. Input P1dB vs. RF Frequency at Various Temperatures, LO = 17 dBm

Figure 44. Input P1dB vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

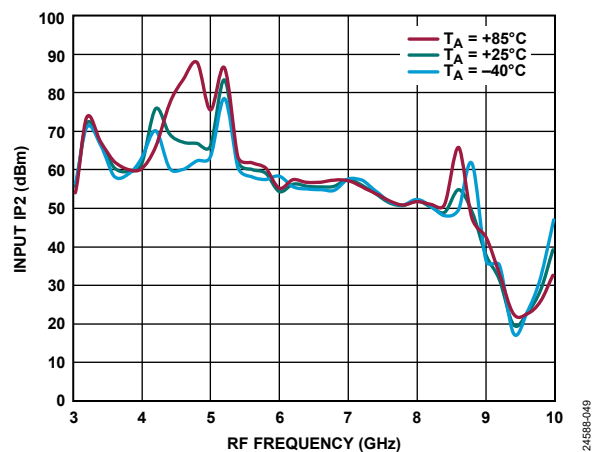


Figure 45. Input IP2 vs. RF Frequency at Various Temperatures,
LO = 17 dBm

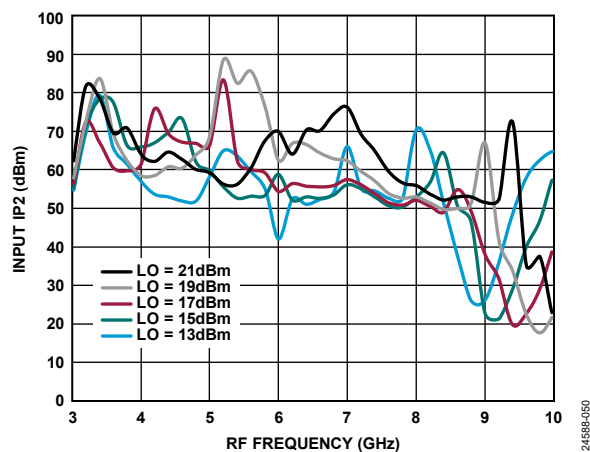


Figure 46. Input IP2 vs. RF Frequency at Various LO Power Levels,
 $T_A = 25^\circ\text{C}$

Upper Sideband (Low-Side LO)

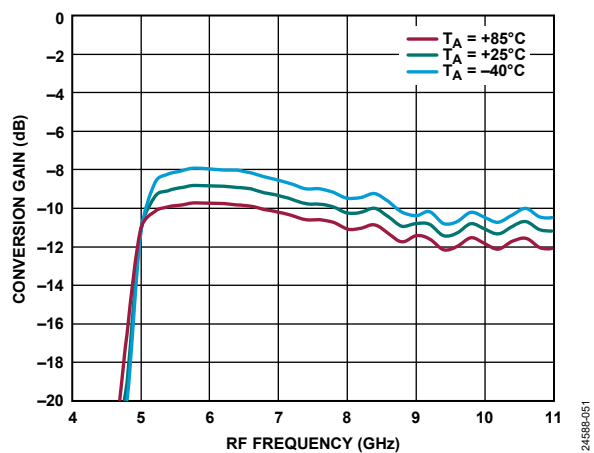


Figure 47. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 17 dBm

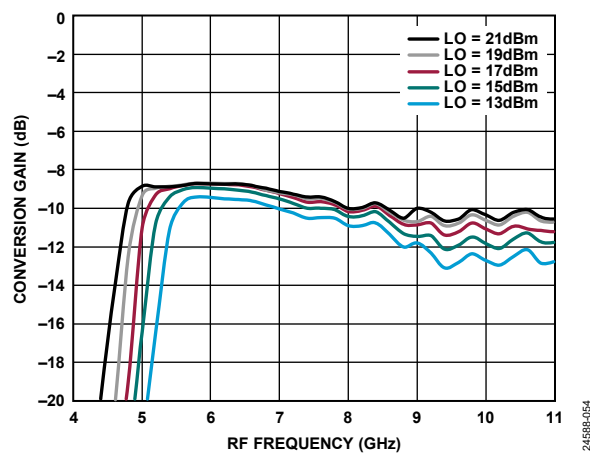
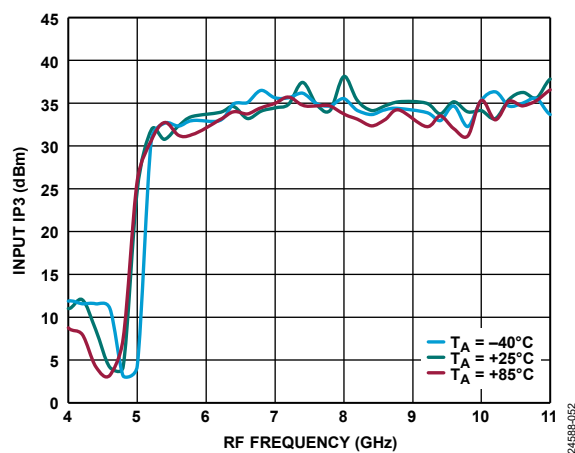
Figure 50. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$ 

Figure 48. Input IP3 vs. RF Frequency at Various Temperatures, LO = 17 dBm

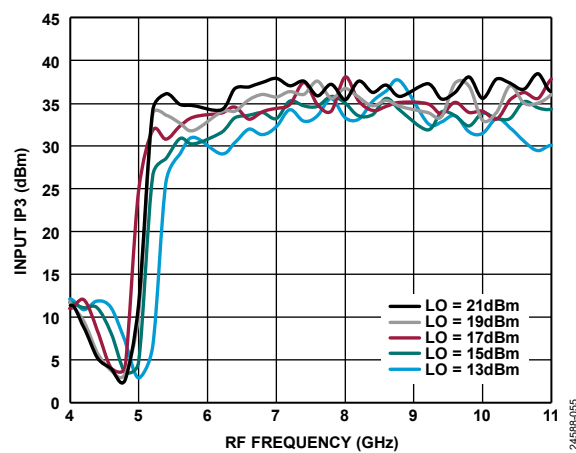
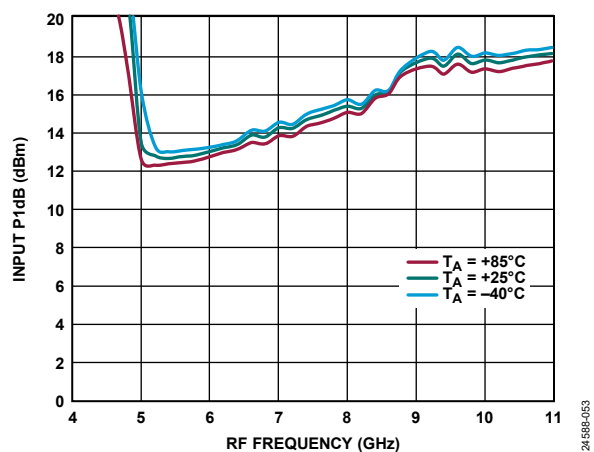
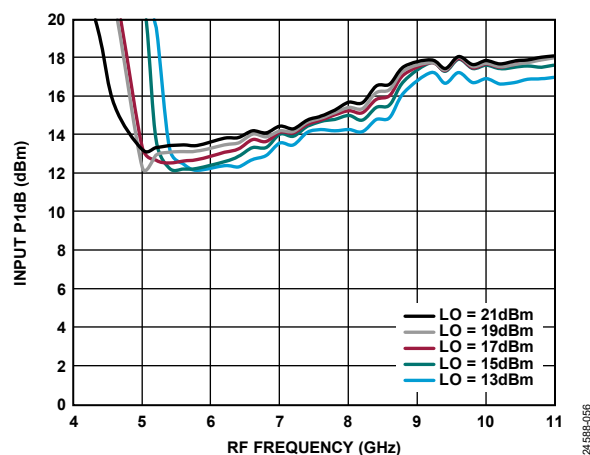
Figure 51. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$ 

Figure 49. Input P1dB vs. RF Frequency at Various Temperatures, LO = 17 dBm

Figure 52. Input P1dB vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

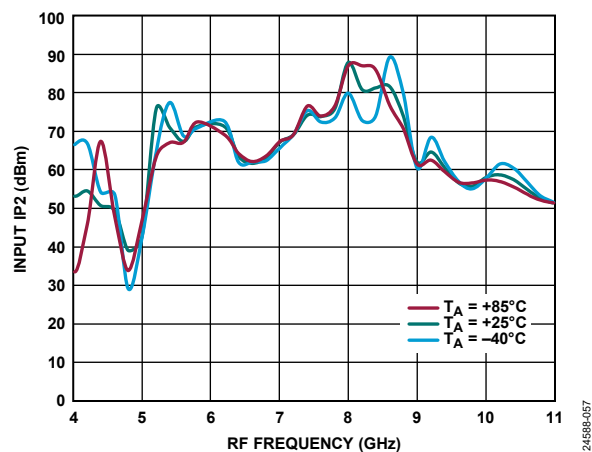


Figure 53. Input IP2 vs RF Frequency at Various Temperatures, LO = 17 dBm

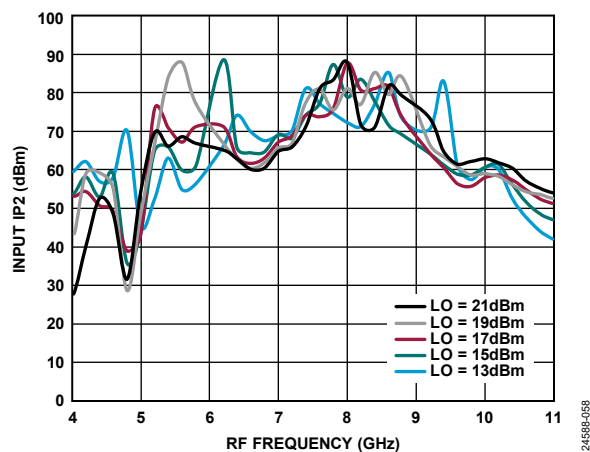


Figure 54. Input IP2 vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

UPCONVERTER PERFORMANCE, IF = 100 MHz

Lower Sideband (High-Side LO)

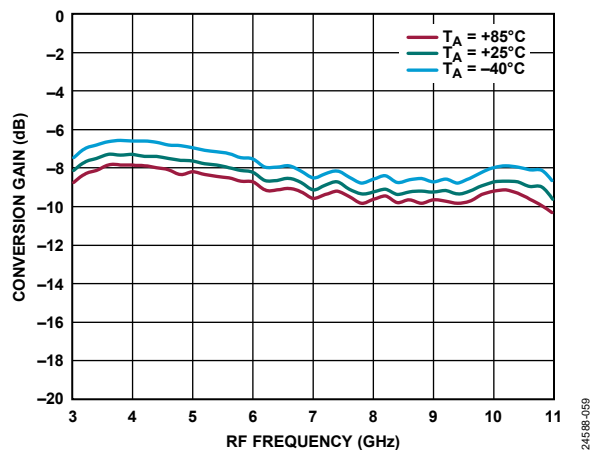


Figure 55. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 17 dBm

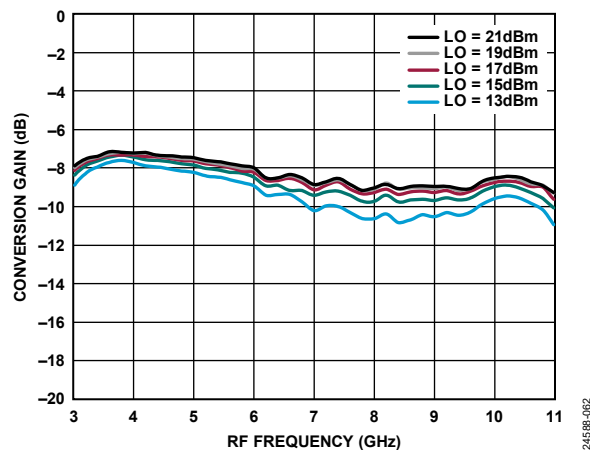


Figure 57. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

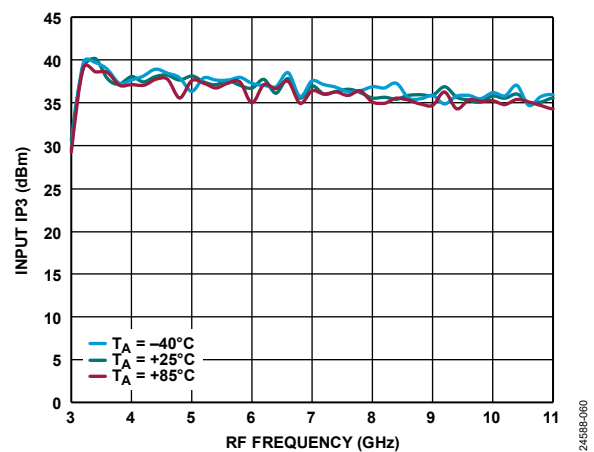


Figure 56. Input IP3 vs. RF Frequency at Various Temperatures, LO = 17 dBm

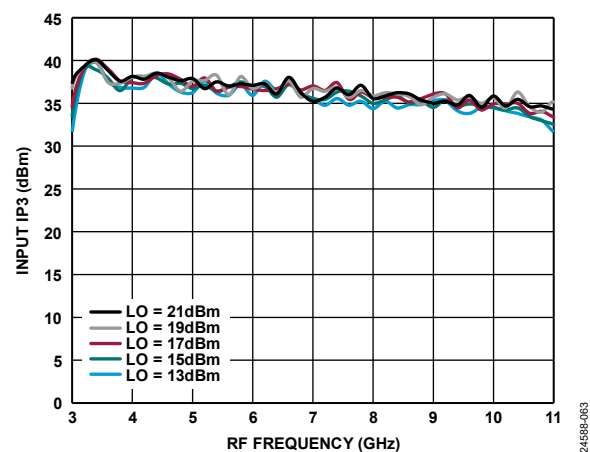


Figure 58. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

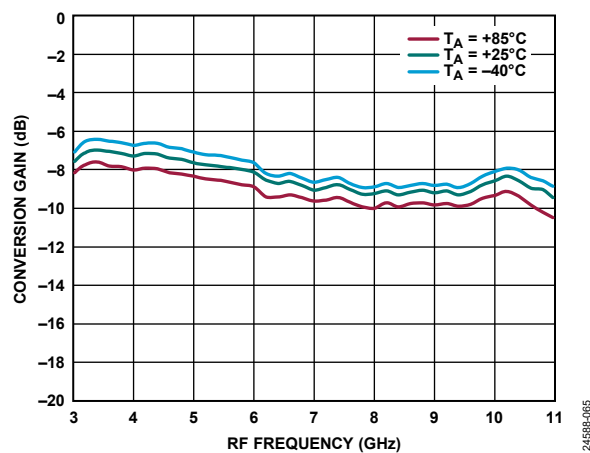
Upper Sideband (Low-Side LO)

Figure 59. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 17 dBm

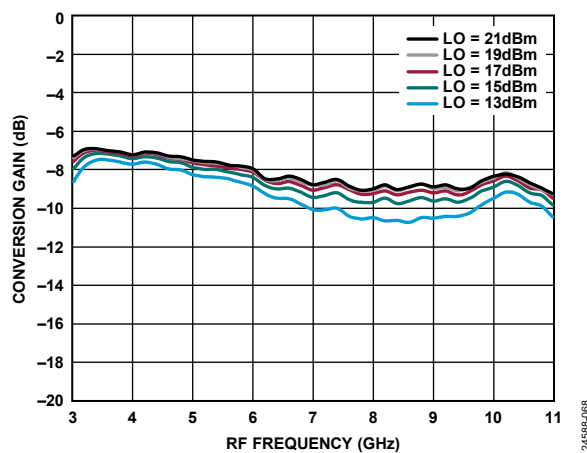


Figure 61. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

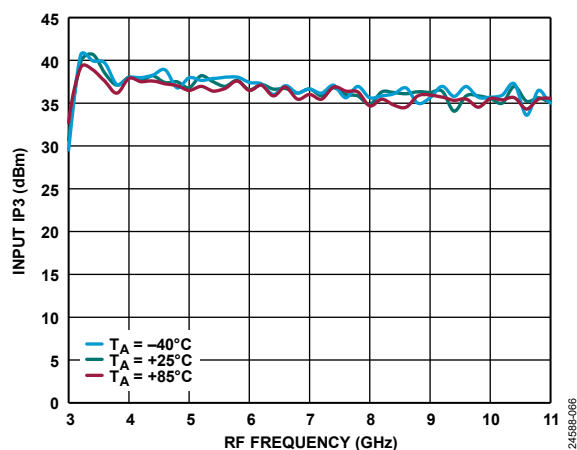


Figure 60. Input IP3 vs. RF Frequency at Various Temperatures, LO = 17 dBm

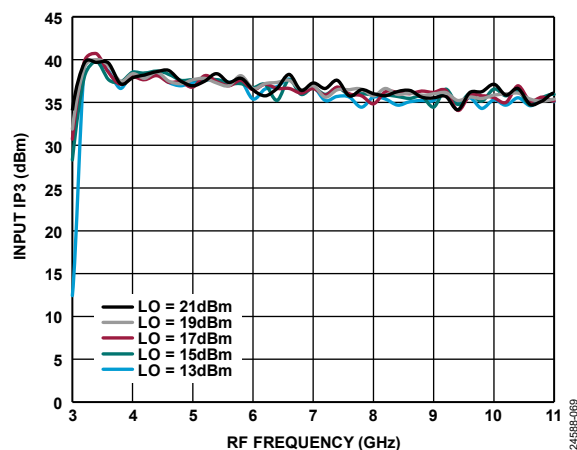


Figure 62. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

UPCONVERTER PERFORMANCE, IF = 1.1 GHz

Lower Sideband (High-Side LO)

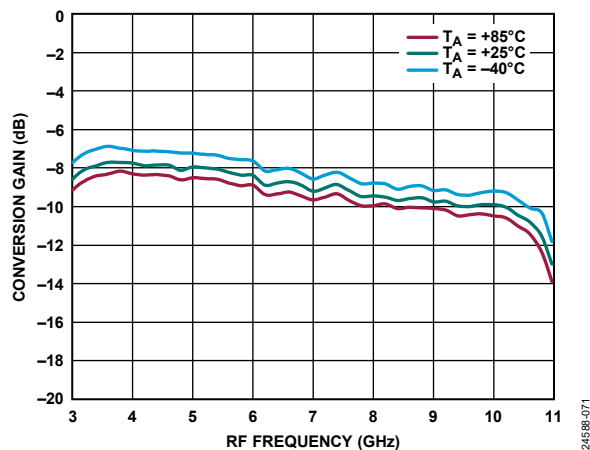


Figure 63. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 17 dBm

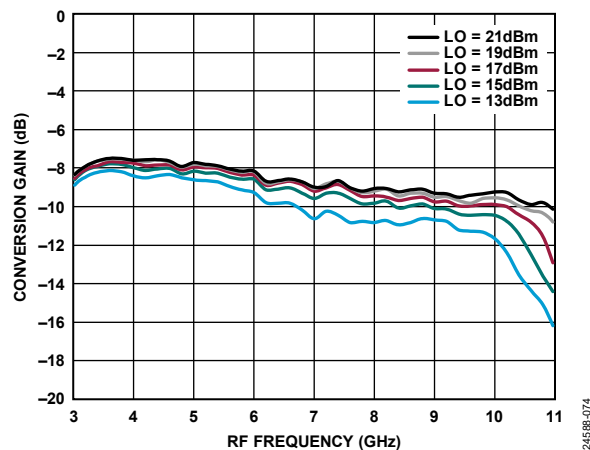
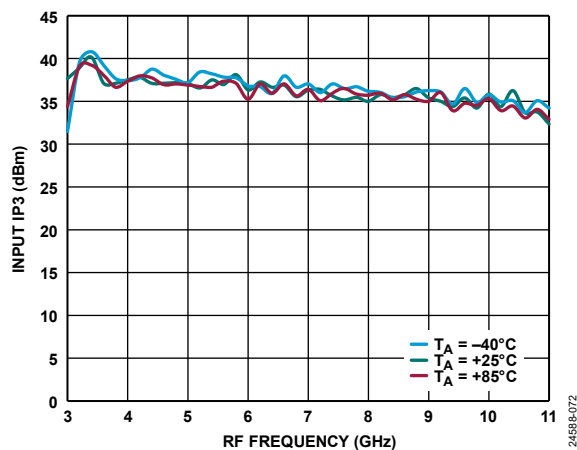
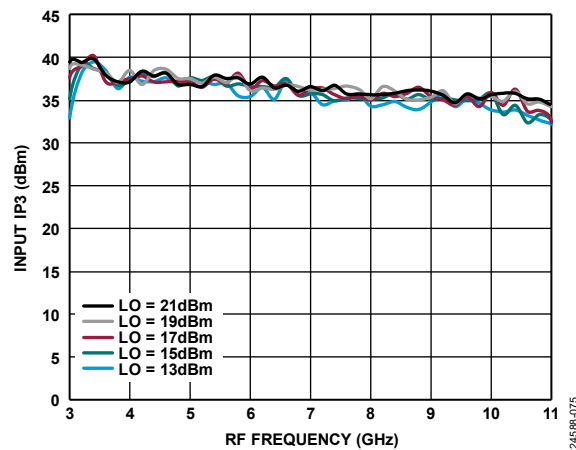
Figure 65. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$ 

Figure 64. Input IP3 vs. RF Frequency at Various Temperatures, LO = 17 dBm

Figure 66. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

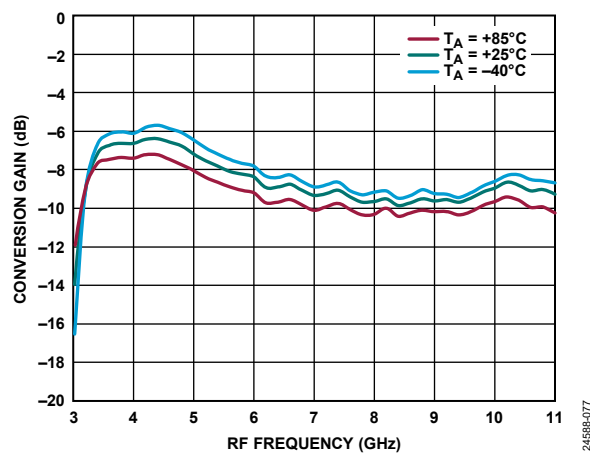
Upper Sideband (Low-Side LO)

Figure 67. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 17 dBm

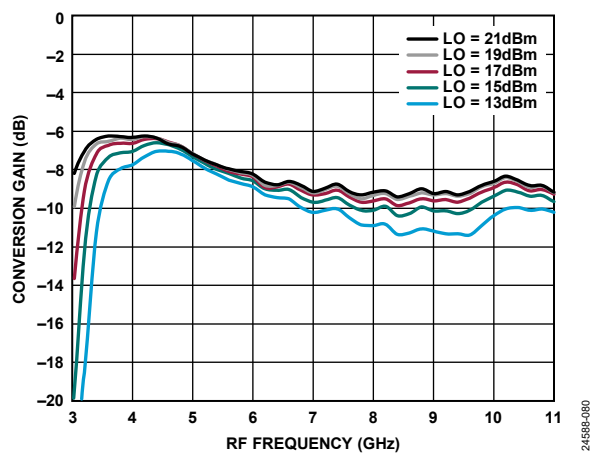


Figure 69. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

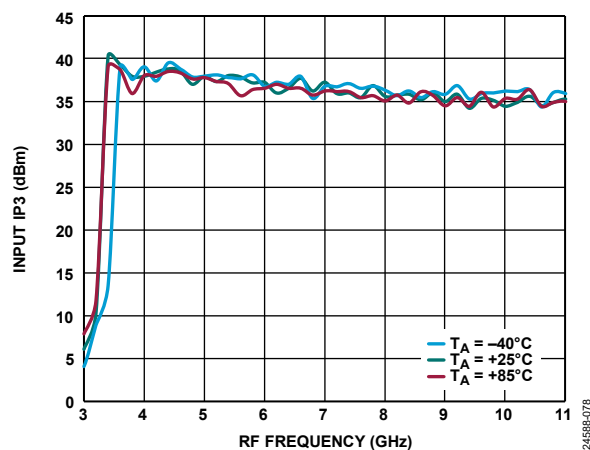


Figure 68. Input IP3 vs. RF Frequency at Various Temperatures, LO = 17 dBm

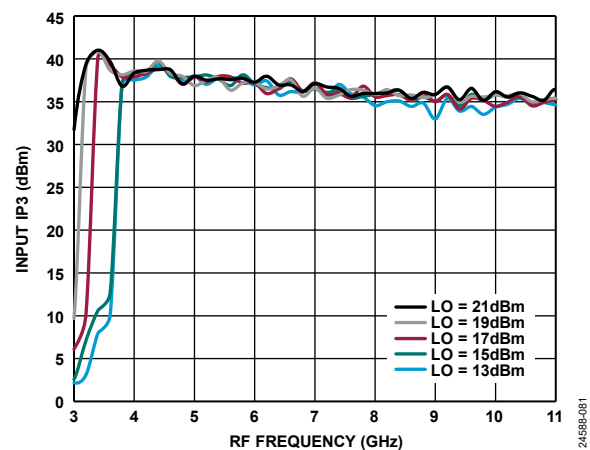


Figure 70. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

UPCONVERTER PERFORMANCE, IF = 3 GHz

Lower Sideband (High-Side LO)

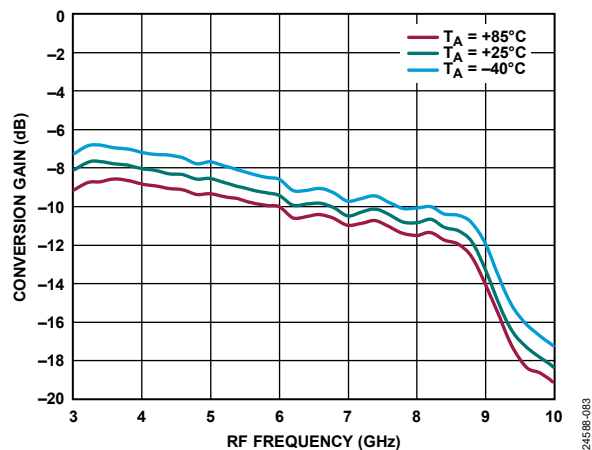


Figure 71. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 17 dBm

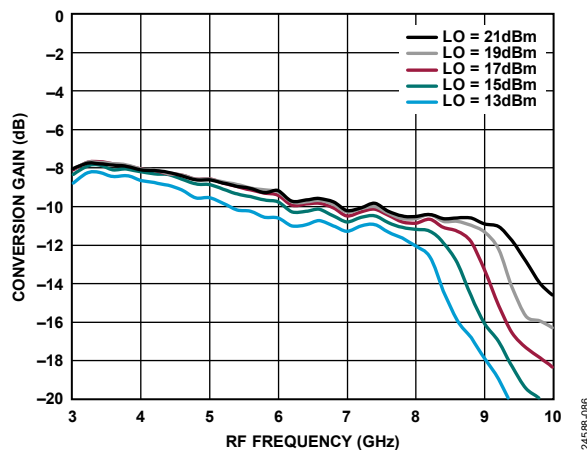


Figure 73. Conversion Gain vs. RF Frequency at Various LO Power Levels, TA = 25°C

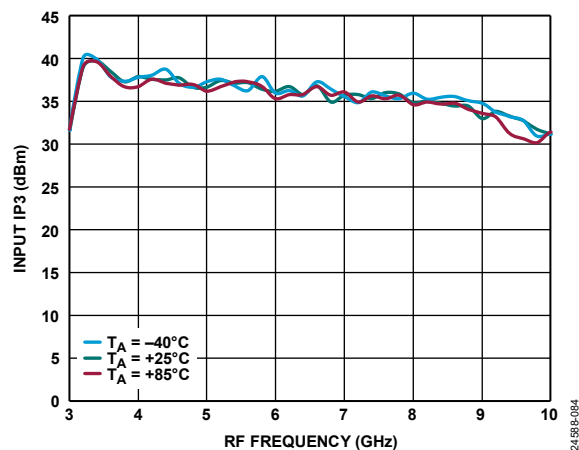


Figure 72. Input IP3 vs. RF Frequency at Various Temperatures, LO = 17 dBm

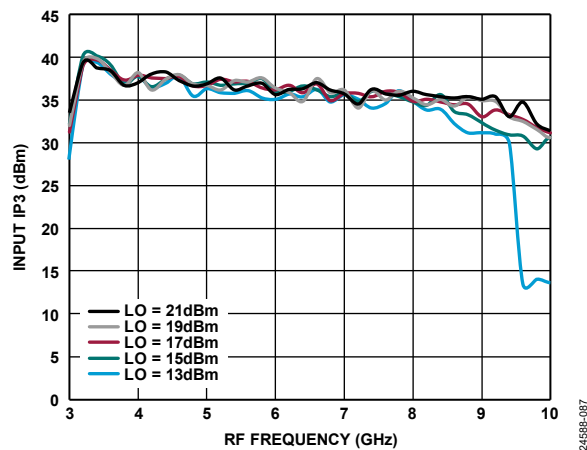


Figure 74. Input IP3 vs. RF Frequency at Various LO Power Levels, TA = 25°C

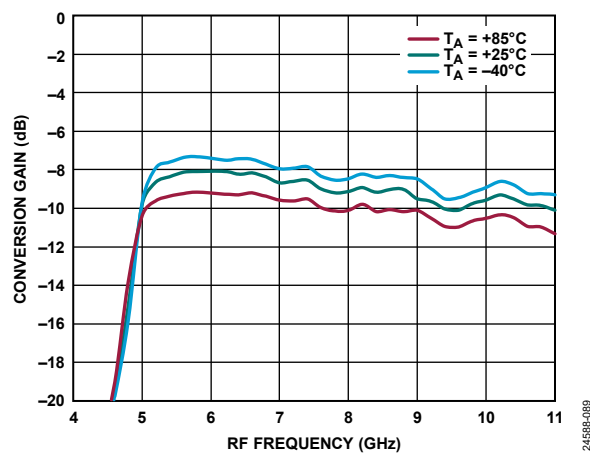
Upper Sideband (Low-Side LO)

Figure 75. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 17 dBm

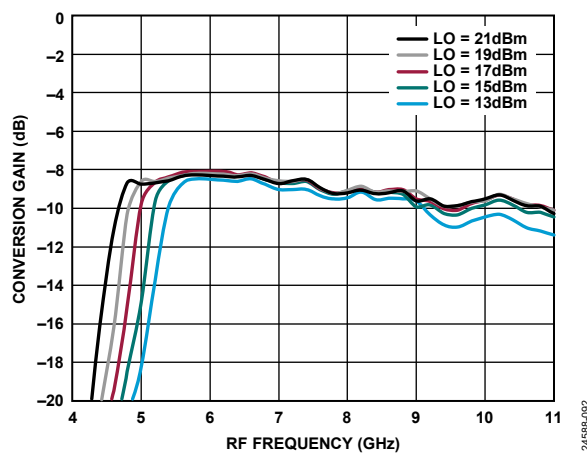


Figure 77. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

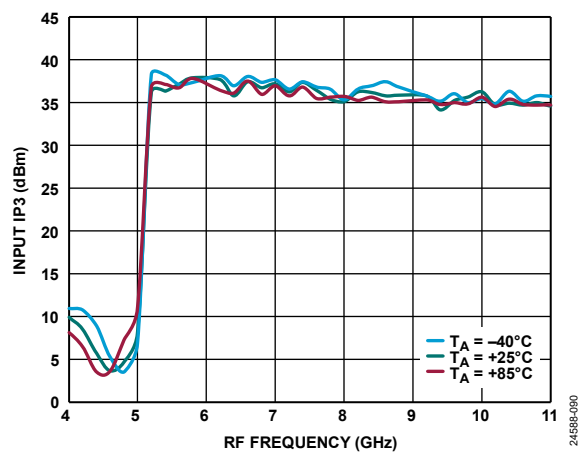


Figure 76. Input IP3 vs. RF Frequency at Various Temperatures, LO = 17 dBm

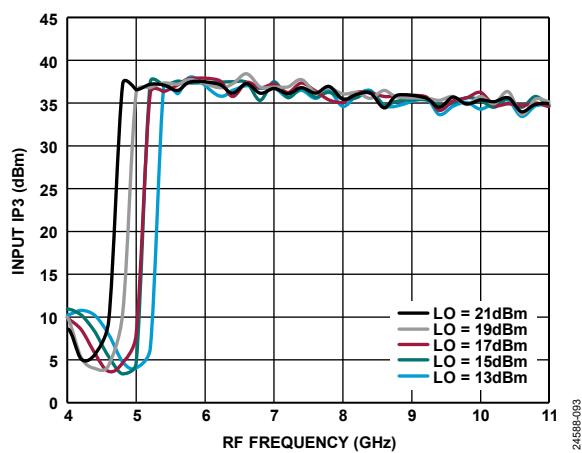


Figure 78. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

ISOLATION AND RETURN LOSS

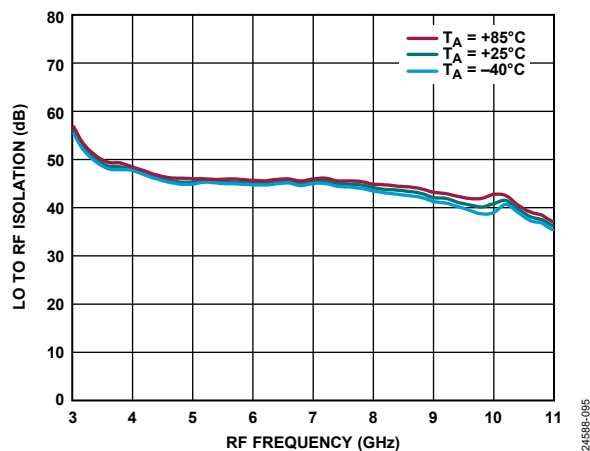


Figure 79. LO to RF Isolation vs. RF Frequency at Various Temperatures, LO = 17 dBm, IF = 100 MHz

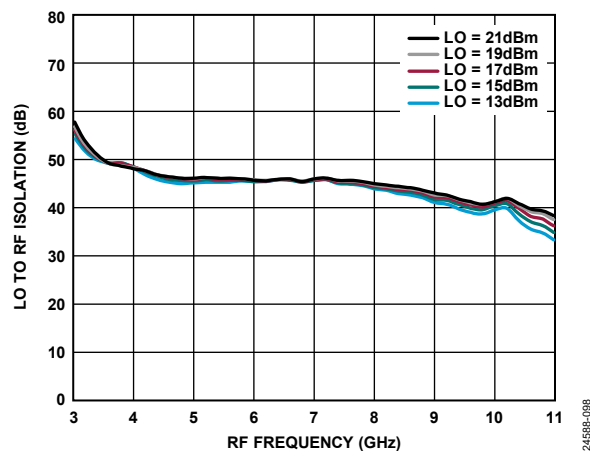


Figure 82. LO to RF Isolation vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$, IF = 100 MHz

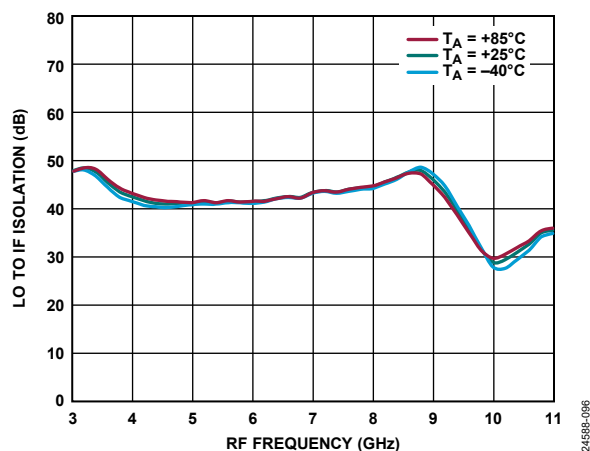


Figure 80. LO to IF Isolation vs. RF Frequency at Various Temperatures, LO = 17 dBm, IF = 100 MHz

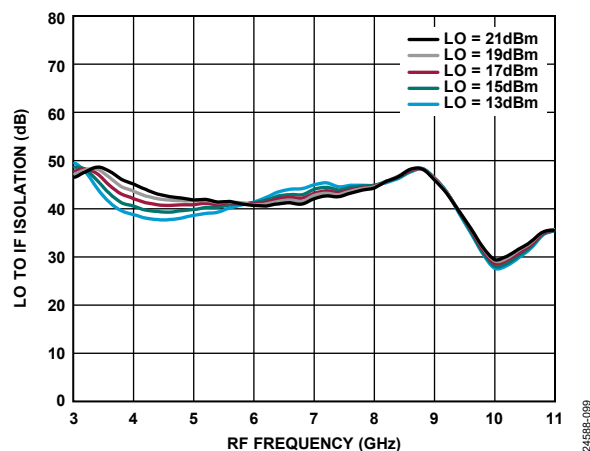


Figure 83. LO to IF Isolation vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$, IF = 100 MHz

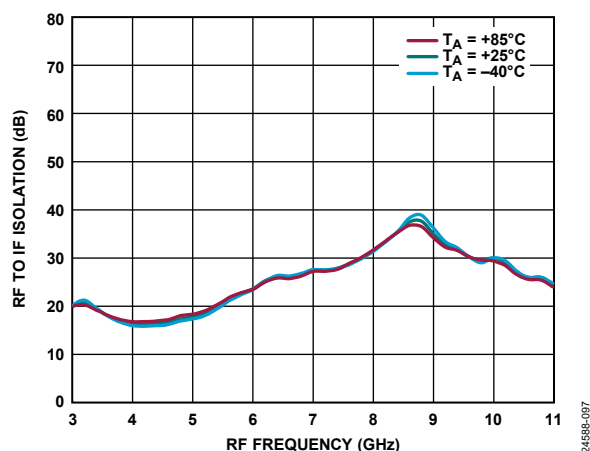


Figure 81. RF to IF Isolation vs. RF Frequency at Various Temperatures, LO = 17 dBm, IF = 100 MHz

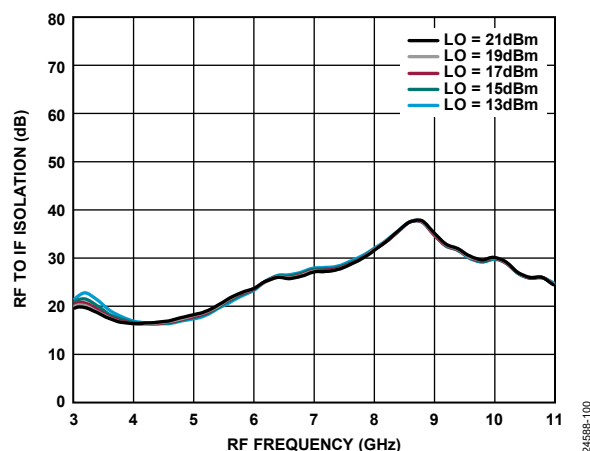


Figure 84. RF to IF Isolation vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$, IF = 100 MHz

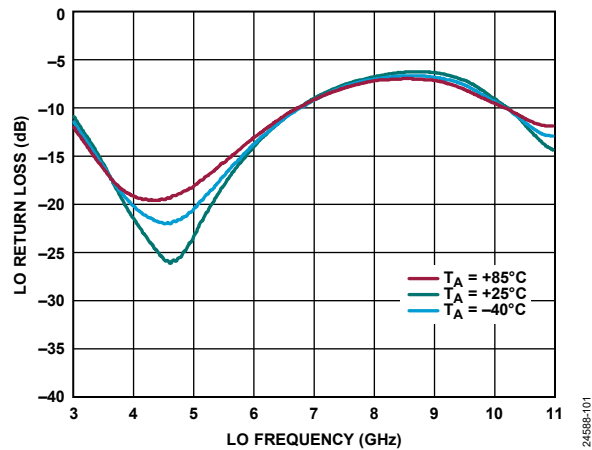


Figure 85. LO Return Loss vs. LO Frequency at Various Temperatures, LO = 17 dBm

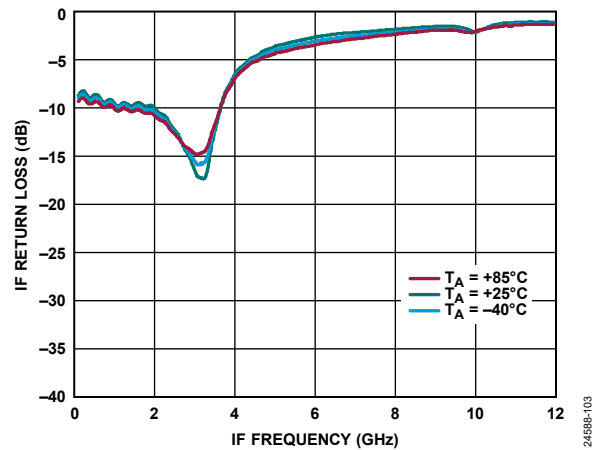


Figure 87. IF Return Loss vs. IF Frequency at Various Temperatures, LO Frequency = 9510 MHz

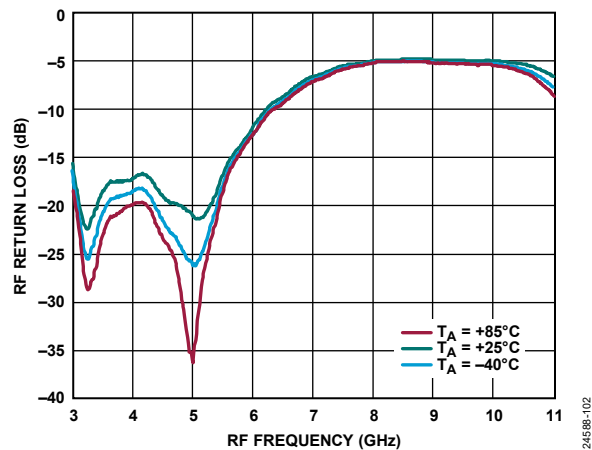


Figure 86. RF Return Loss vs. RF Frequency at Various Temperatures, LO Frequency = 9510 MHz

IF BANDWIDTH—DOWNCONVERTER

Lower Sideband, LO Frequency = 9.51 GHz

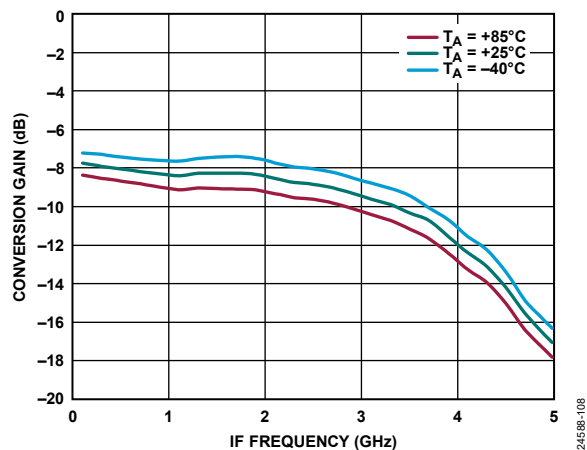


Figure 88. Conversion Gain vs. IF Frequency at Various Temperatures, LO = 17 dBm

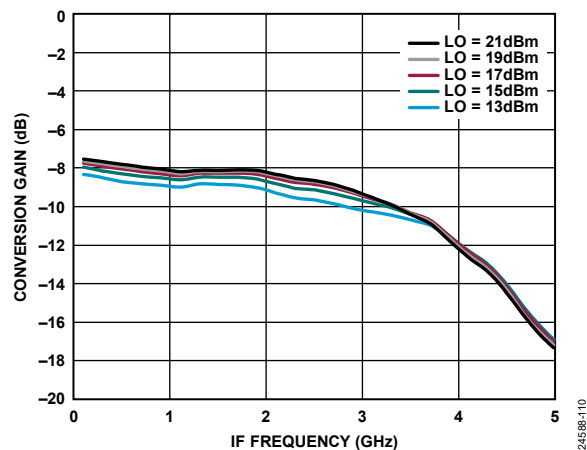
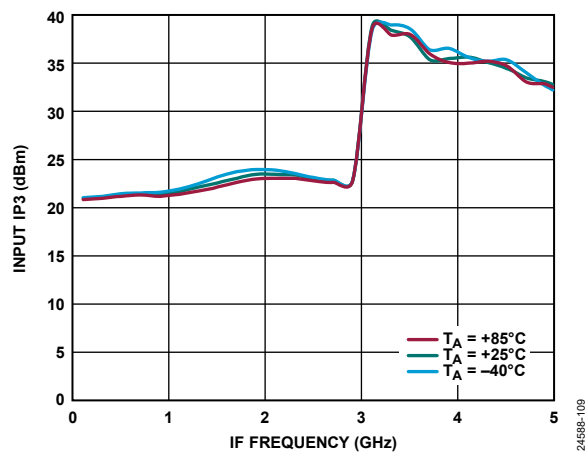
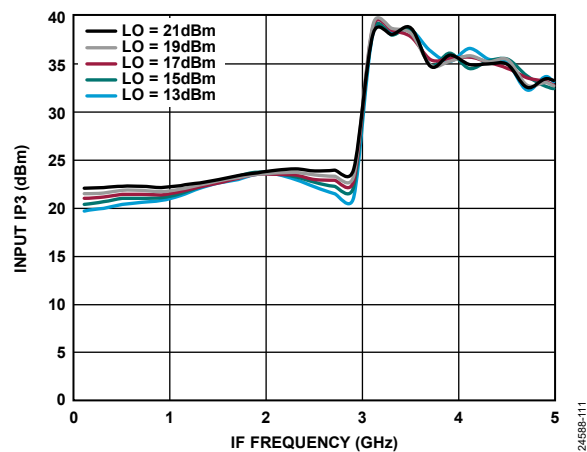
Figure 90. Conversion Gain vs. IF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$ 

Figure 89. Input IP3 vs. IF Frequency at Various Temperatures, LO = 17 dBm

Figure 91. Input IP3 vs. IF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

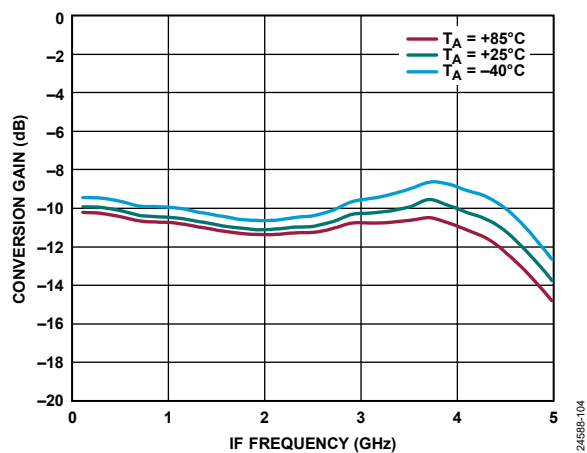
Upper Sideband, LO Frequency = 9.51 GHz

Figure 92. Conversion Gain vs. IF Frequency at Various Temperatures, LO = 17 dBm

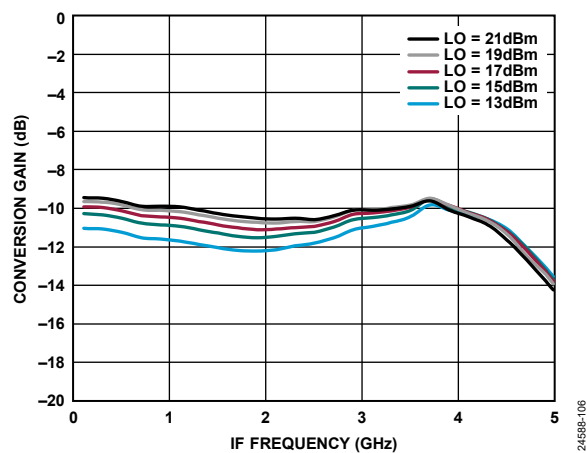
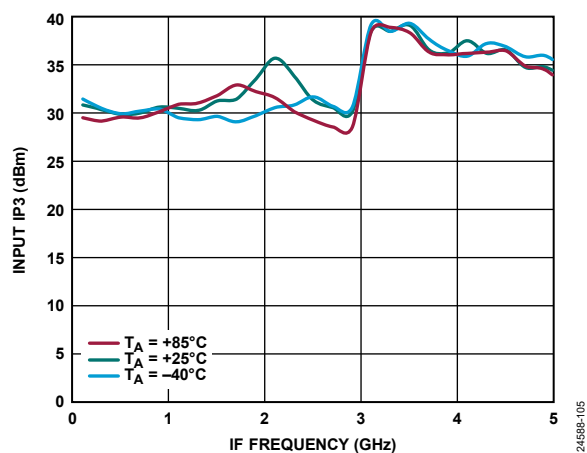
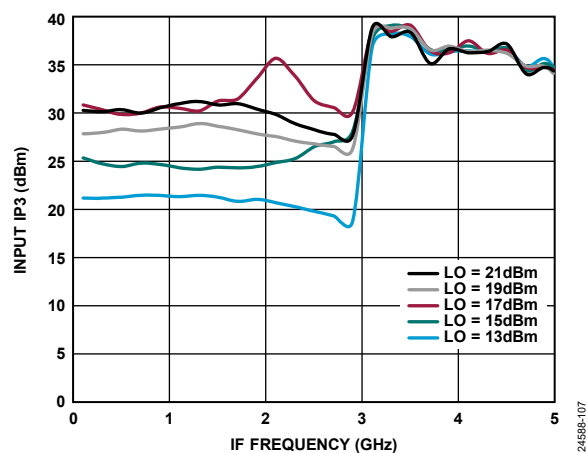
Figure 94. Conversion Gain vs. IF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$ 

Figure 93. Input IP3 vs. IF Frequency at Various Temperatures, LO = 17 dBm

Figure 95. Input IP3 vs. IF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

SPURIOUS PERFORMANCE

LO Harmonics

LO = 17 dBm, and all values in dBc are below the input LO level and measured at the RF port.

Table 5. LO Harmonics at RF

LO Frequency (GHz)	N _{LO} Spur at RF Port (dBc)			
	1	2	3	4
3	54.2	50.4	55.3	61.1
4	48.2	43.8	51.3	70.7
5	46.2	42.3	62.2	81.5
6	45.9	42.4	64.4	71.8
7	46.1	49.6	62.7	80.1
8	44.7	54	63	62.7
9	42.6	58.2	69.6	59.1
10	43	64.5	63.5	56.7

LO = 17 dBm, and all values in dBc are below the input LO level and measured at the IF port.

Table 6. LO Harmonics at IF

LO Frequency (GHz)	N _{LO} Spur at IF Port (dBc)			
	1	2	3	4
3	47.8	67.5	71	86.3
4	42.1	68.3	62	73.8
5	41.2	58.7	73.3	73.5
6	41.7	76.8	77.2	74
7	43.7	62	68.7	75.3
8	44.8	58.9	64.2	74.5
9	44.7	56.9	65.5	67.8
10	29.7	57.3	66.8	69.9

M × N Spurious Outputs

Downconversion, Upper Sideband

Spur values are $(M \times RF) - (N \times LO)$. RF = 3.1 GHz, LO = 3 GHz, RF power = -5 dBm, and LO power = +17 dBm. Mixer spurious products are measured in dBc from the IF output power level.

		N × LO				
		0	1	2	3	4
M × RF	1	12.6	0	31.6	27.1	56.3
	2	86.2	88.5	75.9	77.7	85.2
	3	82.2	84.6	89.5	59.7	69.5
	4	82	81.5	84.9	88.6	92.9

Downconversion, Lower Sideband

Spur values are $(M \times RF) - (N \times LO)$. RF = 2.9 GHz, LO = 3 GHz, RF power = -5 dBm, and LO power = +17 dBm. Mixer spurious products are measured in dBc from the IF output power level.

		N × LO				
		0	1	2	3	4
M × RF	1	0	23.6	27.1	59.1	81.2
	2	81.2	80.2	76.7	87.5	83
	3	82	84.3	62.6	59.2	88.1
	4	81.4	81.7	84.1	87.1	93.6

Upconversion, Upper Sideband

Spur values are $(M \times \text{Input IF (IF}_{IN}) + (N \times LO)$. IF_{IN} = 100 MHz, LO = 6 GHz, IF_{IN} power = -5 dBm, and LO power = +17 dBm. Mixer spurious products are measured in dBc from the RF output power level. N/A means not applicable.

		N × LO				
		0	1	2	3	4
M × IF _{IN}	-4	92.2	84.2	82.6	80.8	75.7
	-3	87.9	85.5	82.3	79.5	73.2
	-2	83.1	84.7	81.1	81	75
	-1	29.1	0	82.9	80.4	74.9
	0	N/A	15.6	12.3	33.1	76.7
	+1	29.1	0	84.5	79.3	77.3
	+2	82.9	84.9	80	80.5	76.9
	+3	87.3	82.4	83.7	78.7	76.0
	+4	95.1	84.3	80.3	77.1	76.9

Upconversion, Lower Sideband

Spur values are $(M \times \text{IF}_{IN}) + (N \times LO)$. IF_{IN} = 100 MHz, LO = 3 GHz, IF_{IN} power = -5 dBm, and LO power = +17 dBm. Mixer spurious products are measured in dBc from the RF output power level. N/A means not applicable.

		N × LO				
		0	1	2	3	4
M × IF	-4	94.1	88.3	84.5	84.7	81.3
	-3	87.6	53.9	85.8	85.8	83
	-2	83.8	62.1	84.4	81.3	81.1
	-1	29.4	0	21.6	11.4	29.5
	0	N/A	24.4	20.1	24.9	30.5
	+1	29.4	0	20.1	12	28.8
	+2	83.2	88.5	83.4	84.1	82.9
	+3	86	87.6	84	82	82
	+4	94.4	91.1	84.1	82.2	81.5

THEORY OF OPERATION

The HMC787A is a general-purpose, double balanced mixer that can be used as an upconverter or downconverter from 3 GHz to 10 GHz. This mixer is fabricated in a GaAs, MESFET process and requires no external components or matching

circuitry. The HMC787A provides high LO to RF and LO to IF isolation due to optimized balun structures and operates with a LO drive level between 13 dBm to 21 dBm.

APPLICATIONS INFORMATION

TYPICAL APPLICATION CIRCUIT

Figure 96 shows the typical application circuit for the [HMC787AG](#). The LO and RF pads are internally ac-coupled. When IF operation is not required until dc, it is recommended to use an ac-coupled capacitor at the IF port. When IF operation to dc is required, do not exceed the IF source and sink currents specified in the Absolute Maximum Ratings section.

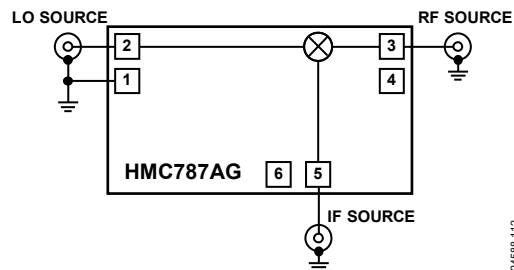


Figure 96. Typical Application Circuit

24598-112

MOUNTING AND BONDING TECHNIQUES

Attach the die directly to the ground plane eutectically or with conductive epoxy. To bring RF to and from the chip, 50 Ω microstrip transmission lines on 0.127 mm (0.005") thick alumina thin film substrates are recommended (see Figure 97).

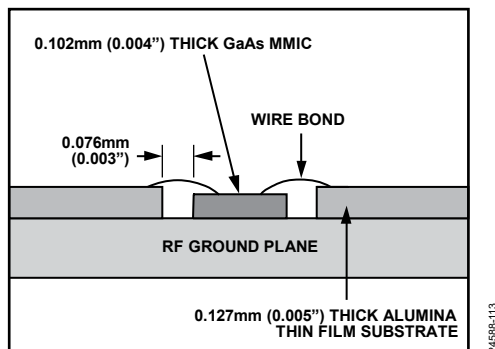


Figure 97. Bonding RF Pads to 5 mil Substrate

If using 0.254 mm (0.010") thick alumina thin film substrates, raise the die 0.150 mm (0.006") so that the surface of the die is coplanar with the surface of the substrate. A way to accomplish this is to attach the 0.102 mm (0.004") thick die to a 0.150 mm (0.006") thick molybdenum heat spreader (moly tab), which is then attached to the ground plane (see Figure 98). To minimize bond wire length, place microstrip substrates as close to the die as possible. Typical die to substrate spacing is 0.076 mm (0.003").

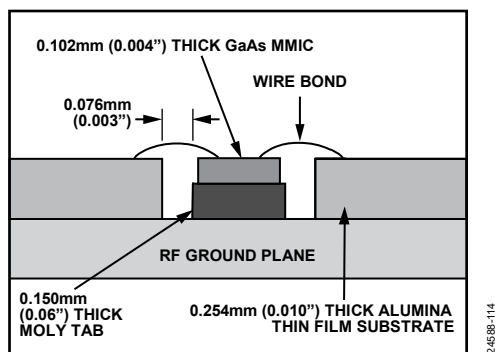


Figure 98. Bonding RF Pads to 10 mil Substrate

HANDLING PRECAUTIONS

To avoid permanent damage to the device, follow the precautions in the following Storage, Cleanliness, Static Sensitivity, Transients, and General Handling sections.

Storage

All bare dice are placed in either waffle- or gel-based ESD protective containers and then sealed in an ESD protective bag for shipment. After opening the sealed ESD protective bag, store all dice in a dry nitrogen environment.

Cleanliness

Handle the chips in a clean environment. Do not attempt to clean the chip using liquid cleaning systems.

Static Sensitivity

Follow ESD precautions to protect against ESD strikes.

Transients

Suppress instrument and bias supply transients while bias is applied. Use shielded signal and bias cables to minimize inductive pickup.

General Handling

Handle the chip along the edges with a vacuum collet or with a sharp pair of bent tweezers. The surface of the chip has fragile air bridges. Do not touch the chip with a vacuum collet, tweezers, or fingers.

MOUNTING

The chip is back metallized and can be die mounted with gold (Au)/tin (Sn) eutectic preforms or with electrically conductive epoxy. The mounting surface must be clean and flat.

Eutectic Die Attach

An 80/20 gold and tin preform is recommended with a work surface temperature of 255°C and a tool temperature of 265°C. When hot 90/10 nitrogen(N)/hydrogen (H) gas is applied, the tool tip temperature must be 290°C. Do not expose the chip to a temperature greater than 320°C for more than 20 seconds. No more than 3 sec of scrubbing is required for attachment.

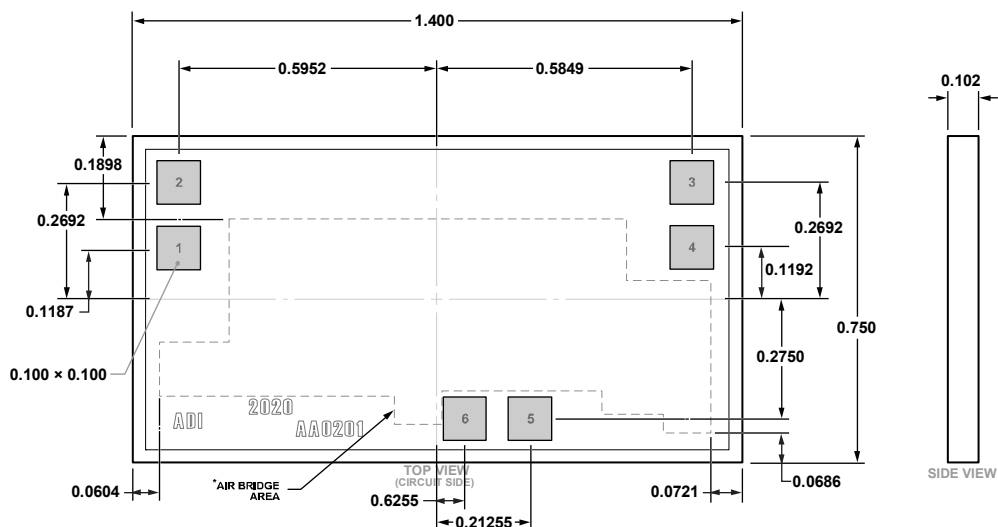
Epoxy Die Attach

Apply a minimum amount of epoxy to the mounting surface so that a thin epoxy fillet is observed around the perimeter of the chip when the chip is placed into position. Cure epoxy per the schedule of the manufacturer.

WIRE BONDING

Ball or wedge bond with 0.025 mm (0.00098") diameter pure gold wire is recommended. Thermosonic wire bonding with a nominal stage temperature of 150°C and a ball bonding force of 40 grams to 50 grams, or a wedge bonding force of 18 grams to 22 grams, is recommended. Use the minimum level of ultrasonic energy to achieve reliable wire bonds. Wire bonds must begin on the chip and terminate on the package or substrate. All bonds must be as short as possible <0.31 mm (0.01220").

OUTLINE DIMENSIONS



*This die utilizes fragile air bridges. Any pickup tools used must not contact this area.

Figure 99. 6-Pad Bare Die [CHIP]
(C-6-15)

Dimensions shown in millimeters

ORDERING GUIDE

Model ¹	Temperature Range	Package Description	Package Option
HMC787AG	−40°C to +85°C	6-Pad Bare Die [CHIP]	C-6-15
HMC787AG-SX	−40°C to +85°C	6-Pad Bare Die [CHIP]	C-6-15

¹ The HMC787AG and HMC787AG-SX are RoHS compliant parts.

Mouser Electronics

Authorized Distributor

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[HMC787ALC3BTR](#) [HMC787ALC3BTR-R5](#) [HMC787ALC3B](#) [EV1HMC787ALC3B](#) [HMC787AG](#) [HMC787AG-SX](#)